AN EXAMINATION OF THE EFFECT OF THE VARIABLE PART OF BLADE
CIRCULATION ON THE INDUCED POWER REQUIREMENTS OF A TWO-
BLADED ROTOR IN FORWARD FLIGHT BY AN ELECTROMAGNETIC
ANALOG

A THESIS

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the Faculty of the Graduate Division
by
William L. Corley, Sr.

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of the Requirements for the Degree
Master of Science in Aeronautical Engineering

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AN EXAMINATION OF THE EFFECT OF THE VARIABLE PART OF BLADE CIRCULATION ON THE INDUCED POWER REQUIREMENTS OF A TWO-BLADED ROTOR IN FORWARD FLIGHT BY AN ELECTROMAGNETIC ANALOG

Approved:

Date Approved by Chairman: September 5, 1961
I would like to express my sincere appreciation to Dr. Robin B. Gray for his assistance in conducting this study. Also, gratitude is extended to Professor Walter Castles, Jr. for selection of the topic. To Dr. Thomas W. Jackson of the Engineering Experiment Station and Professor Howard L. Durham, I would like to express my appreciation for reviewing and commenting on this thesis. Thanks are extended to Dr. Frank O. Nottingham of the School of Electrical Engineering for designing the current transformers used in the electromagnetic analog.
# TABLE OF CONTENTS

<table>
<thead>
<tr>
<th>Section</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ACKNOWLEDGMENTS</td>
<td>ii</td>
</tr>
<tr>
<td>LIST OF FIGURES</td>
<td>iv</td>
</tr>
<tr>
<td>LIST OF SYMBOLS</td>
<td>v</td>
</tr>
<tr>
<td>SUMMARY</td>
<td>vii</td>
</tr>
<tr>
<td>CHAPTER</td>
<td></td>
</tr>
<tr>
<td>I. INTRODUCTION</td>
<td>1</td>
</tr>
<tr>
<td>II. APPARATUS AND EXPERIMENTIAL PROCEDURE</td>
<td>2</td>
</tr>
<tr>
<td>Primary Field Coil (Wake Model)</td>
<td></td>
</tr>
<tr>
<td>Reference Coil</td>
<td></td>
</tr>
<tr>
<td>Search Coil</td>
<td></td>
</tr>
<tr>
<td>Amplifier and Output Meter</td>
<td></td>
</tr>
<tr>
<td>Power Supply</td>
<td></td>
</tr>
<tr>
<td>Field Survey Procedure</td>
<td></td>
</tr>
<tr>
<td>Reduction of Data</td>
<td></td>
</tr>
<tr>
<td>III. RESULTS</td>
<td>6</td>
</tr>
<tr>
<td>IV. CONCLUSIONS</td>
<td>7</td>
</tr>
<tr>
<td>FIGURES</td>
<td>8</td>
</tr>
<tr>
<td>APPENDIX</td>
<td>21</td>
</tr>
<tr>
<td>BIBLIOGRAPHY</td>
<td>24</td>
</tr>
<tr>
<td>Figure</td>
<td>Description</td>
</tr>
<tr>
<td>--------</td>
<td>-----------------------------------------------------------------------------</td>
</tr>
<tr>
<td>1.</td>
<td>Helical and Pie-Shaped Vortices</td>
</tr>
<tr>
<td>2.</td>
<td>Primary Coil (Wake Model)</td>
</tr>
<tr>
<td>3.</td>
<td>Reference Coil</td>
</tr>
<tr>
<td>4.</td>
<td>Search Coil</td>
</tr>
<tr>
<td>5.</td>
<td>Amplifier and Meter</td>
</tr>
<tr>
<td>6.</td>
<td>Wake Model Circuit Diagram</td>
</tr>
<tr>
<td>7.</td>
<td>Electromagnetic Analog Wiring Diagram</td>
</tr>
<tr>
<td>8.</td>
<td>Variation of Non-Dimensional Normal Component of Induced Velocity Across Diameter of Two-Bladed Rotor for Constant and Variable Blade Circulation</td>
</tr>
<tr>
<td>(a)</td>
<td>For $\psi = 90^\circ - 270^\circ$</td>
</tr>
<tr>
<td>(b)</td>
<td>For $\psi = 60^\circ - 240^\circ$</td>
</tr>
<tr>
<td>(c)</td>
<td>For $\psi = 30^\circ - 210^\circ$</td>
</tr>
<tr>
<td>(d)</td>
<td>For $\psi = 0^\circ - 180^\circ$</td>
</tr>
<tr>
<td>(e)</td>
<td>For $\psi = 330^\circ - 150^\circ$</td>
</tr>
<tr>
<td>(f)</td>
<td>For $\psi = 300^\circ - 120^\circ$</td>
</tr>
</tbody>
</table>
LIST OF SYMBOLS

A  area of the rotor disc
B  magnetic flux density
b  number of blades of the rotor

\( C_T \)  rotor thrust coefficient  \( C_T = \frac{T}{\rho \pi B b^4} \)

\( C_{P_I} \)  induced power coefficient

\( M_T \)  rotor thrust moment

MR  meter reading, decibels

N  subscript for a value at the reference point

n  number of turns of wire on the reference coil

P  subscript for a value at an arbitrary point \( P \)

\( P_I \)  induced power

R  rotor radius

\( R_c \)  radius of reference coil

\( R_m \)  wire model rotor radius

r  radial distance from rotor axis

T  rotor thrust

\( U_T \)  tangential component of resultant velocity at blade element,

\[ U_T = 2 \pi r + \mu R \sin \psi \]

V  forward velocity of rotor

\( V_1 \)  component of induced velocity normal to the rotor plane

w  component of induced velocity normal to the rotor plane

x  non-dimensional radial distance along blade, \( r/R \)

\( \alpha \)  rotor tip-path angle of attack

\( \Gamma \)  total blade circulation
constant part of blade circulation

\( \Gamma_0 \)

variable part of blade circulation

\( \Gamma_1 \)

non-dimensionalised circulation strength,

\[ \gamma = \frac{\Gamma}{\omega T L \times R^2} \]

rotor tip speed ratio,

\[ \mu = \frac{\omega \times \cos \chi}{U R} \]

mass density of air

\( \rho \)

wake skew angle, angle between axis of wake and rotor tip-path plane

\( \chi \)

azimuth angle of blade feathering axis measured from downwind position and in direction of blade rotation

\( \psi \)

rotor angular velocity

\( \Omega \)
SUMMARY

A wire model of the wake vortices of a rotor blade in forward flight at a blade tip forward speed ratio, \( \mu \), of 0.3, and wake skew angle, \( \chi \), of 90° was constructed to simulate constant circulation across the rotor blade and lateral harmonic variation in circulation with each 30° azimuth position of the blade.

Electrical current, analogous to the non-dimensionalized values of constant and variable blade circulation, was passed through the wire model and the component of the magnetic field strength normal to the tip-path plane was measured across the rotor diameter for each 30° blade azimuth angle. These measured components of magnetic field strength were converted to induced velocities, non-dimensionalized in terms of the rotor radius and blade circulation. From symmetry and the principle of superposition, the effect of the second blade was obtained to give the total value for a two-bladed rotor. This procedure was repeated using only the vortices shed at the blade root and tip to obtain the effect of only constant circulation across the blade.

The measurements were used in numerically integrating the induced power coefficient equation to obtain a comparison of the effect of assuming lateral harmonic variations in blade circulation, as opposed to assuming only constant circulation, on the induced power requirements of a two-bladed rotor in forward flight at a blade tip forward speed ratio of 0.3.

The results indicate that by assuming only constant circulation, the induced power requirements are underestimated by 16.5 per cent.
CHAPTER I

INTRODUCTION

The wake pattern of a lifting rotor operating in forward flight has been shown to consist of continuous skewed helical vortex filaments caused by the variation in blade vortex strength along the blade radius and closed, pie-shaped vortex filaments caused by the harmonic change of blade bound vortex strength with azimuth angle. An illustration of these filaments, obtained from Reference 1, is given in Fig. 1. Consideration of the pie-shaped vortex filaments, or the variable part of blade circulation, in computing the induced velocity of a rotor has largely been neglected in theory. Also, little consideration has been given to the effect of these pie-shaped vortices on the variations in blade bending stresses across the rotor.

In this study an examination of the effect of the lateral harmonic variations of blade circulation at 30° azimuth angles will be made by using the results obtained from an electromagnetic analog to compare the induced power requirements of a two-bladed rotor in forward flight at a rotor tip forward speed ratio of 0.3 with that computed for a similar rotor assuming only constant blade circulation.
CHAPTER II

APPARATUS AND EXPERIMENTAL PROCEDURE

The equipment used for the experimental portion of this thesis consisted of the following components:

1. The primary coil (wire model of the helical and pie-shaped vortex filaments);
2. The reference coil;
3. The secondary coil (search coil);
4. The amplifier and output meter;
5. The power supply.

Primary Coil.—The primary coil, shown in Fig. 2, was constructed of No. 17 gage, Brown and Sharpe copper wire to represent the helical and pie-shaped vortex filaments. Since the wake skew angle, \( \chi \), was assumed to be 90°, the wake extended directly behind the rotor. The rotor radius was ten inches, with the blade extending from the 0.2 radius. Since uniform blade bound vortex strength was assumed along the radius, helical vortex filaments were shed at the blade tip and 0.2 radius. These helical vortex filaments, or constant part of blade bound vortex strength, were represented by a continuous circuit with an arbitrary constant current of 3.0 amperes. The pie-shaped vortex filament strengths, or variable part of the blade bound vortex strength, were calculated over \( 30^\circ \) intervals and were represented with a current equivalent to the ratio of the variable and constant blade circulation multiplied by the 3.0 amperes. Since the variable part of the blade
bound vortex strength is a function of \( \sin \psi \), the pie-shaped filaments could be represented by three circuits, arranged to give a reduction in field strength from \( \psi = 0^\circ \) to \( \psi = 180^\circ \) and an addition of field strength from \( \psi = 180^\circ \) to \( \psi = 360^\circ \). An illustration of these circuits is given in Fig. 6. The correct current was obtained in the circuits by using current transformers that were wound in a ratio of the constant to the variable part of the blade bound vortex strength. In order to simplify the model, the vortices from only one blade were constructed and the results for two blades were obtained by superposing the computed non-dimensionalized values of induced velocity, obtained from measurements taken across the diameter of the rotor.

Reference Coil.--The reference coil consisted of nine closely spaced turns of No. 17 gage wire wrapped on a plexiglas ring. A switching arrangement was used so that the coil could be placed in any of the four circuits. This allowed a check to be made of the current ratios. The meter readings were referenced to the magnetic field strength at the center of the reference coil while in the 3.0 ampere circuit. The reference coil is shown in Fig. 3.

Search Coil.--The search coil consisted of 1000 turns of No. 40, Brown and Sharpe, copper wire wound on a plexiglas core which was mounted on a plexiglas support as shown in Fig. 4. The coil had a diameter of 0.35 inches to the center of the wire bundle which had a cross section approximately 0.09 inches square. A co-axial cable was used to connect the search coil to the amplifier.

Amplifier and Output Meter.--The pickup circuit included a commercial standing wave indicator in addition to the search coil and co-axial cable
described above. This indicator, which had a maximum sensitivity of 0.1 microvolt, consisted of an indicating meter, a high-gain 400 cycle fixed frequency amplifier with a calibrated gain control covering a 60 decibel range, and a narrow 400 cycle band-pass-filter network which had a sharp cutoff at 400 \(+\) 5 cycles per second. The input impedance of the amplifier was 200,000 ohms, sufficient to eliminate any calibration of the meter scale. The amplifier was placed in a separate room from the model to eliminate any appreciable magnetic coupling from these coils. The standing wave indicator is shown in Fig. 5.

Power Supply.--The power supply was a 400 cycle, 115 volt, aircraft alternator driven by a synchronous motor. A variable resistor was adjusted to allow 3.0 amperes of current to flow in the circuit and enough capacitance was placed in parallel with the circuit to obtain a power factor of unity. The power supply and current transformers were in a different room from the wake model to minimize direct magnetic coupling effects. A wiring diagram of the system is shown in Fig. 7.

Field Survey Procedure.--The search coil was placed directly over the wake model and was aligned so as to pick up only the normal component of the magnetic field strength. With the rotor blade initially in the position of $\Psi = 90^\circ$, measurements were made from $r/R = 1.10$ to $r/R = 0.00$. Then by measuring from $r/R = 0.00$ to $r/R = 1.10$ for $\Psi = 270^\circ$, the effect of the blade at $\Psi = 90^\circ$ on the blade at $\Psi = 270^\circ$ could be superposed on the measurements taken when the blade was actually in the $\Psi = 270^\circ$ position. The blade position was changed by cutting out 30° of the circuit and re-wiring for each survey.

Twelve surveys were conducted to obtain the non-dimensional nor-
mal component of induced velocity over the entire rotor. Prior to each survey the equipment was allowed a thirty minute warm-up period and the search coil was centered in the plane of the reference coil to obtain an adjusted reference meter reading.

Reduction of Data.—The output meter was calibrated in decibels and it was necessary to convert the meter readings into equivalent values of the non-dimensional normal component of induced velocity. This conversion was accomplished through the following formulas, derived in Reference 2.

\[
\frac{B_p}{B_N} = \frac{\text{antilog } 0.1 (\text{MR})_P}{\text{antilog } 0.1 (\text{MR})_N}
\]

\[
\frac{\nu R}{R} = \frac{n(R_m)B_p}{2 R_c B_N}
\]

where

- \( B_p \): magnetic flux density at point under consideration
- \( B_N \): magnetic flux density at center of reference coil
- \( \text{MR} \): meter reading, decibels
- \( w \): component of induced velocity normal to the rotor disc
- \( R \): rotor radius
- \( n \): number of turns on reference coil
- \( R_m \): wire model rotor radius
- \( R_c \): radius of reference coil

The direction of the induced velocity was determined from the vortex geometry and a study of the trends in the experimental data.
CHAPTER III

RESULTS

Figures 8(a) through 8(f) show the non-dimensionalized values of the normal component of induced velocity across the rotor diameter of a two-bladed rotor in forward flight at a rotor tip ratio of 0.3. These values were obtained assuming lateral harmonic variations in blade circulation with each 30 degree azimuth angle and also by assuming constant blade circulation. A comparison of the effect of assuming harmonic variations in blade circulation on the induced power requirements of a two-bladed rotor was obtained by numerically integrating the induced power coefficient equation

\[ C_{p1} = \frac{b}{4\pi R^3} \int_0^{2\pi} \int_0^1 V_1 F(x, \psi) \, dx \, d\psi \]

\[ F(x, \psi) = (2 - 3\mu \sin \psi)(x + \mu \sin \psi) \]

utilizing the data from Fig. 8(a) through 8(f). The induced power coefficient equation was obtained from Appendix C of Reference 1.

The results of this numerical integration indicate that if only constant circulation is assumed, the induced power required by a two-bladed rotor in forward flight at a blade tip forward speed ratio of 0.3 will be underestimated by 16.5 per cent.
CHAPTER IV

CONCLUSIONS

The accuracy of the experimental measurements of this study is not known. The greatest source of error is probably in assuming a wake skew angle of 90 degrees. The electromagnetic analog method has certain inherent inaccuracies as noted in Reference 2:

1. Small variations in primary coil current and frequency;
2. Search coil positioning errors and associated meter reading errors;
3. Inaccuracies in the meter and amplifier calibration;
4. Small distortion in the portion of the model magnetic field of interest arising from the laboratory structure;
5. Errors in plotting and drawing the graphs.

There is no known mathematical solution of this study to use for comparative purposes, but it is stated in Reference 1 that an approximate computation of the induced power required for rotors having one, two, three, and four blades and operating at \( \mu = 0.2 \) and \( \mu = 0.4 \), respectively, indicates that previous methods underestimate the induced power by an amount which increases with \( \mu \) and decreases with the number of blades. For the case mentioned above, the per cent underestimation in induced power required is about 39%, 17%, 6%, and 6% for \( \mu = 0.2 \) and 71%, 18% and 15% for \( \mu = 0.4 \). The results of this study support the approximate computations of Reference 1.
Fig. 1 Helical and Pie-Shaped Vortices
Figure 2. Primary Coil (Wake Model).
Figure 3. Reference Coil.
Figure 4. Search Coil.
Figure 5. Amplifier and Meter.
Fig. 6 Wake Model Circuit Diagram
Fig. 7 Electromagnetic Analog Wiring Diagram
Fig. 8 Variation of Non-Dimensional Normal Component of Induced Velocity Across Rotor for Constant and Variable Blade Circulation
Fig. 8 (continued)

(b) For $\psi = 60^\circ - 240^\circ$
Fig. 8 (continued)

(c) For $\psi = 30^\circ - 210^\circ$
Fig. 8 (continued)

(d) For $\psi = 0^\circ - 180^\circ$

For $\psi = 0^\circ - 180^\circ$, the graph shows the variation of $\frac{\tau}{R}$ with $r/R$. The graph includes two lines: one for Variable and another for Constant. The graph indicates significant fluctuations in the variable with respect to $r/R$.
Fig. 8 (continued)

(e) For $\psi = 330^\circ - 150^\circ$
Variable

Constant

(f) For $\psi = 300^\circ - 120^\circ$

Fig. 8 (continued)
APPENDIX

CALCULATION OF THE NON-DIMENSIONAL RATIO OF VARIABLE TO CONSTANT CIRCULATION STRENGTH

Assuming constant circulation along the blade radius and lateral harmonic variations with azimuth angle

\[ \Gamma = \Gamma_0 + \Gamma_1 \sin \psi \]

\[ dT = \rho U_T \Gamma dr \]

\[ T = \frac{b}{2} \oint \oint_{0.2R} R \left( \partial r + \mu \omega R \sin \psi \right) \left( \Gamma_0 + \Gamma_1 \sin \psi \right) d\psi dr \]

\[ T = b \rho \frac{\Delta R^2}{2} \left( 0.96 \Gamma_0 + 0.8 \mu \Gamma_1 \right) \]

\[ C_T = \frac{T}{2\pi \rho R^2 \left( \Gamma_0 + \Gamma_1 \sin \psi \right)} \]

\[ C_T = \frac{b}{2} \frac{0.96 \Gamma_0 + 0.8 \mu \Gamma_1}{\pi \beta R^2} \]

\[ M_T = \oint_{0.2R} \rho \left( \partial r + \mu \omega R \sin \psi \right) \left( \Gamma_0 + \Gamma_1 \sin \psi \right) r dr \]

\[ M_T = \rho \omega R^3 \left\{ \left[ 0.3306 \Gamma_0 \right] + \left[ \frac{0.96 \mu \Gamma_0}{2} + \frac{0.992 \Gamma_1}{3} \right] \sin \psi \right. \]

\[ + \left[ \frac{0.96 \mu}{3} \Gamma_1 \right] \sin^2 \psi \]
The first harmonic of the thrust moment must be zero.

\[
\frac{0.992}{3} r_1 = \frac{0.96 \mu \Gamma}{a}
\]

\[
\Gamma_1 = -1.4516 \mu \Gamma_o
\]

\[
C_T = \frac{b(0.96 \Gamma_o - 1.1613 \mu^2 \Gamma_o)}{\pi a R^2}
\]

\[
\Gamma_o = \frac{2 C_T \Omega \pi R^2}{b(0.96 - 1.1613 \mu^2)}
\]

\[
\Gamma_1 = -1.4516 \mu \frac{2 C_T \Omega \pi R^2}{b(0.96 - 1.1613 \mu^2)}
\]

\[
\gamma = \frac{\Gamma}{2 C_T \Omega \pi R^2} = \frac{1}{b(0.96 - 1.1613 \mu^2)} \left[ 1 - 1.4516 \mu \sin \psi \right]
\]

\[
\gamma_{\text{constant}} = \frac{1}{b(0.96 - 1.1613 \mu^2)}
\]

\[
\gamma_{\text{variable}} = \frac{1.4516 \mu \sin \psi}{b(0.96 - 1.1613 \mu^2)}
\]

For \( b = 2 \) and \( \mu = 0.3 \)

\[
\gamma_{\text{constant}} = 0.58445
\]
\( \gamma \) constant is circuit 4 and was represented with 3.0 amperes of current.

\[
\gamma_{\text{variable}} = -0.2545 \sin \psi
\]

\[
\frac{d \gamma_{\text{variable}}}{d \psi} = \text{sheet strength} = -0.2545 \cos \psi
\]

\[
\gamma_{15^\circ} = \int_{0}^{30^\circ} -0.2545 \cos \psi \, d \psi
\]

Integrating this expression every 30 degrees for 360 degrees yields the non-dimensional values of the variable part of blade circulation.

The ratios of variable to constant circulation and the representative currents are given below.

<table>
<thead>
<tr>
<th>Circuit No.</th>
<th>( \gamma_{\text{var}} / \gamma_{\text{const}} )</th>
<th>Current (Amperes)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>0.2176</td>
<td>0.653</td>
</tr>
<tr>
<td>2</td>
<td>0.1595</td>
<td>0.468</td>
</tr>
<tr>
<td>3</td>
<td>0.0583</td>
<td>0.175</td>
</tr>
<tr>
<td>4</td>
<td>1.000</td>
<td>3.000</td>
</tr>
</tbody>
</table>

The desired current was not obtained in all circuits but time did not allow a re-wiring of the current transformers to obtain the exact ratios.

The ratios of \( \gamma_{\text{var}} / \gamma_{\text{const}} \) obtained follow:

<table>
<thead>
<tr>
<th>Circuit No.</th>
<th>( \gamma_{\text{var}} / \gamma_{\text{const}} )</th>
<th>Per Cent Error</th>
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</thead>
<tbody>
<tr>
<td>1</td>
<td>0.1738</td>
<td>20.0</td>
</tr>
<tr>
<td>2</td>
<td>0.1349</td>
<td>15.5</td>
</tr>
<tr>
<td>3</td>
<td>0.0575</td>
<td>1.2</td>
</tr>
<tr>
<td>4</td>
<td>1.000</td>
<td>0.0</td>
</tr>
</tbody>
</table>

